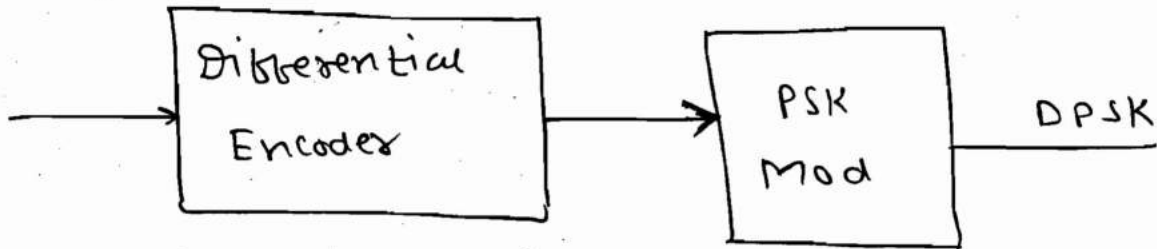


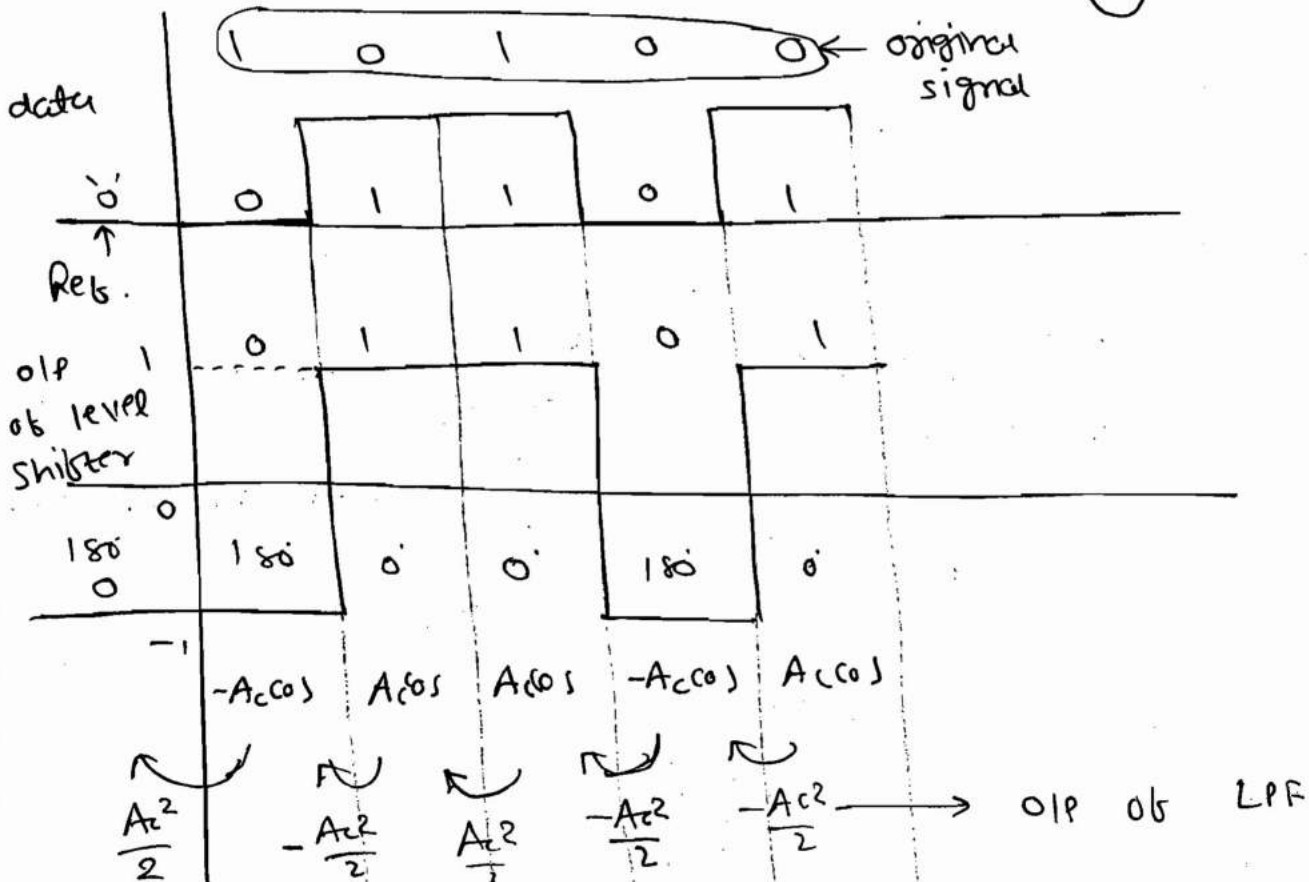
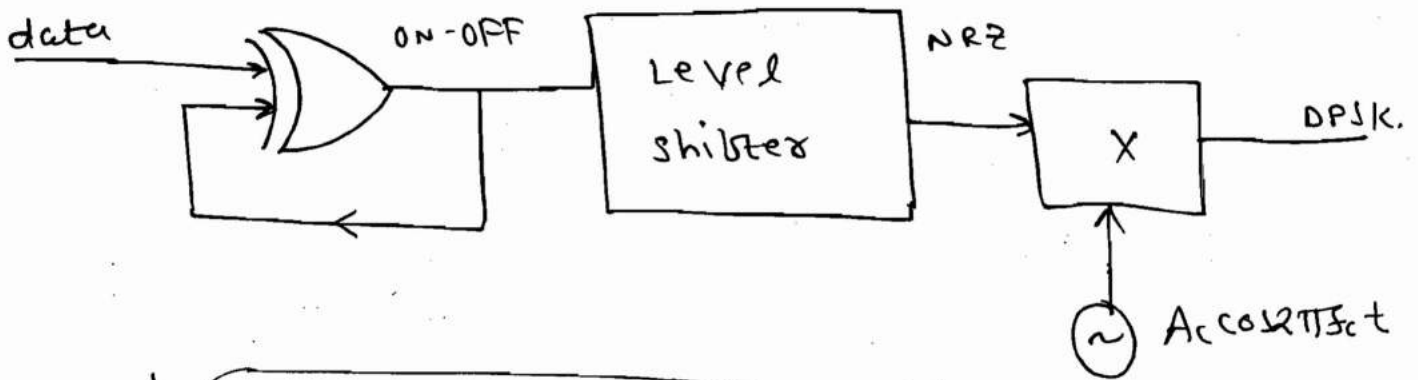
☆ Differential PSK:

$\Rightarrow$ The binary data is differentially encoded and applied to a PSK modulator to generate the DPSK signal.

⇒

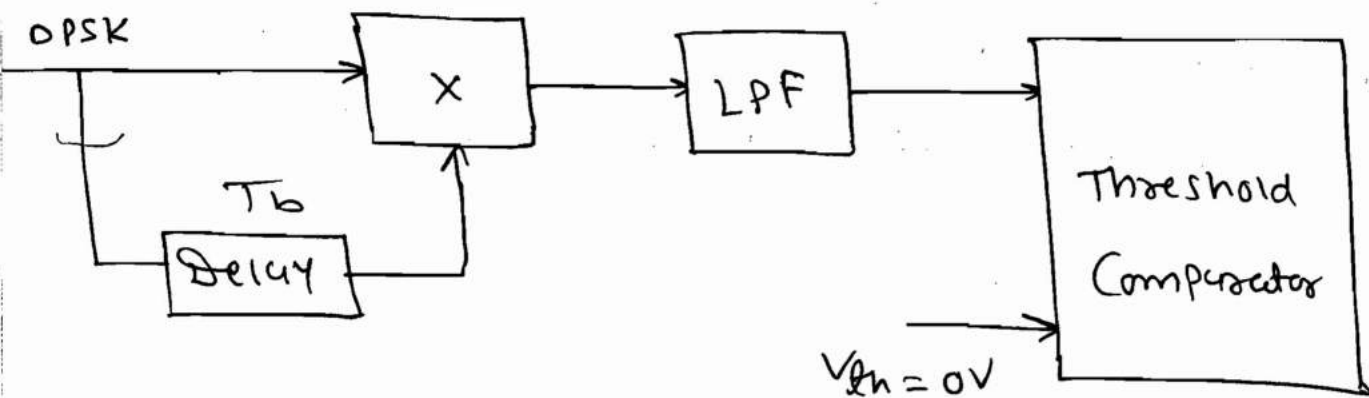


⇒



* Demodulation of DPSK:

⇒



⇒ Local oscillator is not required in DPSK. Synchronization is not required betⁿ the Tx & Rx.

⇒

001001	001001
'1' → 011011	'0' → 100100
π 0 0 π 0 0	0 π π 0 π π

☆ M-ary Signaling:

⇒ Binary signaling

M=2 → 1 → f_1
 0 → f_2 BFSK

→ 1 → ϕ_1
 0 → ϕ_2 BPSK

M=4 QFSK

→ M=4 QPSK

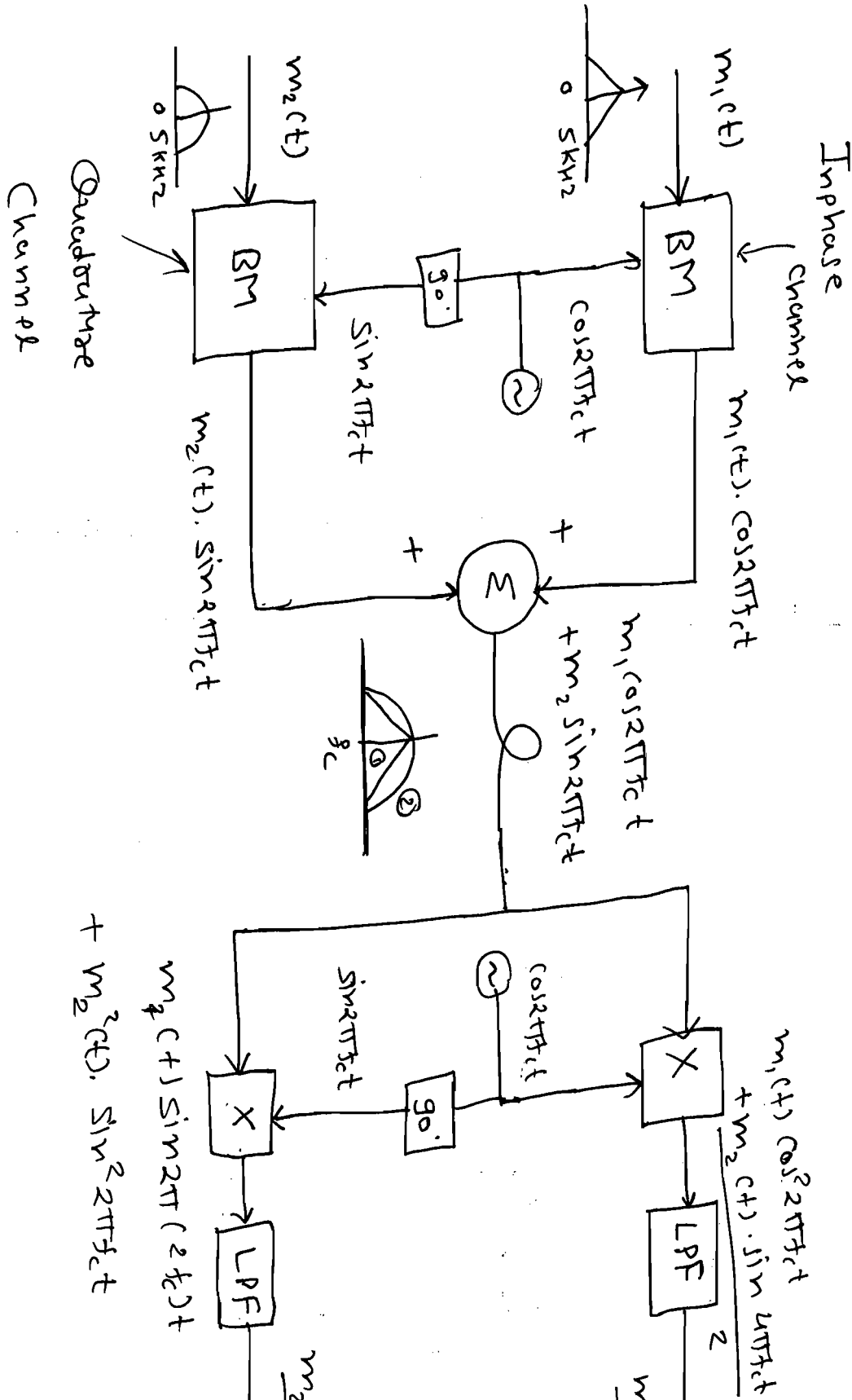
— ϕ_1
 — ϕ_2

M=8 → 8PSK

— ϕ_1
 — ϕ_2
 —
 —
 —
 —

⇒ In M-ary Signalling M - discrete voltage levels are transmitted into free space using digital communication.

* Quadrature Carrier Multiplexing:



⇒ It is also called as orthogonal freq. division multiplexing.

$$\int f_1(t) \cdot f_2(t) dt = 0$$

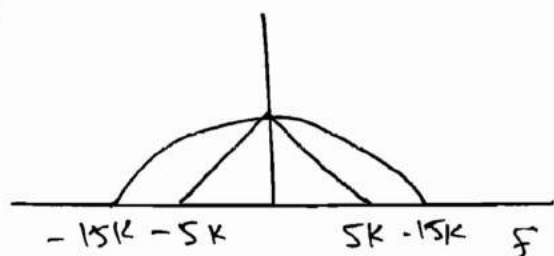
⇒ It is used in CDMA mobile phone.

Q Consider a multiplexed signal

$m_1(t) \cos 2\pi f_c t + m_2(t) \sin 2\pi f_c t$. $m_1(t)$ is band-limited to 10 kHz and $m_2(t)$ is band-limited to 15 kHz. Determine BW of the multiplexed signal:

Solⁿ:

⇒

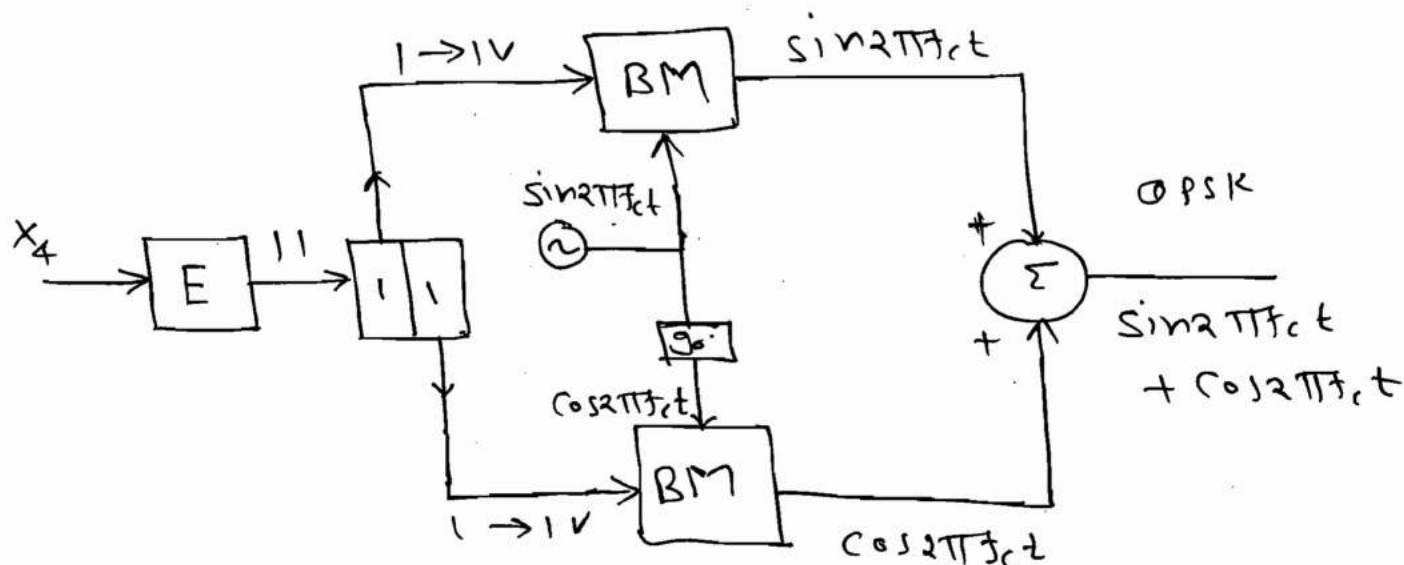


⇒ BW = 2 x highest freq.

$$BW = 30 \text{ kHz}$$

* Quadrature Phase Shift Keying: (QPSK).

⇒



$$\Rightarrow X_1 \text{ --- } 00 \rightarrow \phi_1 = -135^\circ.$$

$$X_2 \text{ --- } 01 \rightarrow \phi_2 = 135^\circ.$$

$$X_3 \text{ --- } 10 \rightarrow \phi_3 = -45^\circ.$$

$$X_4 \text{ --- } 11 \rightarrow \phi_4 = 45^\circ.$$

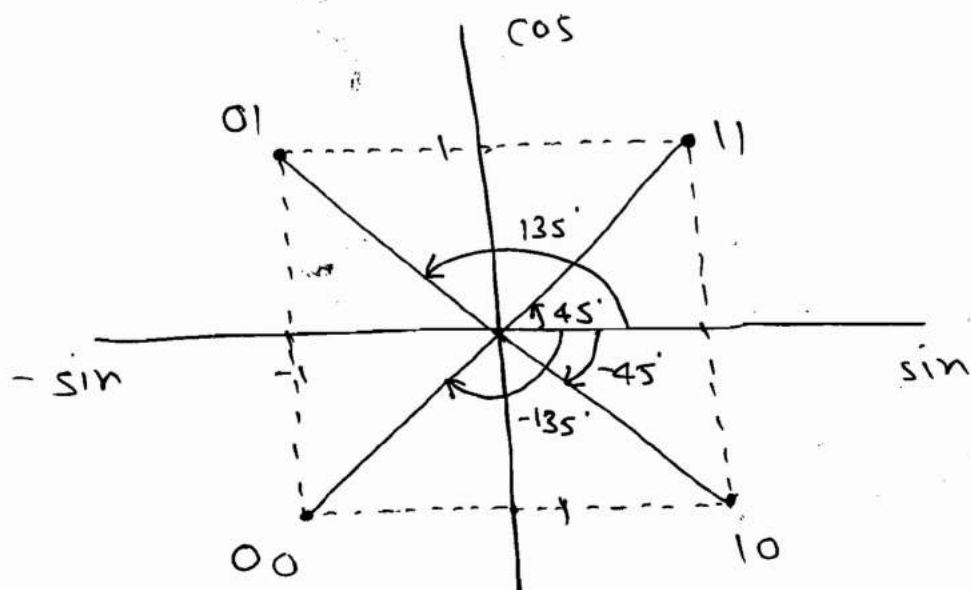
$$\begin{aligned} \Rightarrow X_1 \rightarrow 00 &\rightarrow -\sin 2\pi f_c t - \cos 2\pi f_c t \\ &= \sqrt{2} (\sin(2\pi f_c t + 225^\circ)) \\ &= \sqrt{2} (\sin(2\pi f_c t - 135^\circ)). \end{aligned}$$

$$\begin{aligned} \Rightarrow X_2 \rightarrow 01 &\rightarrow -\sin 2\pi f_c t + \cos 2\pi f_c t \\ &= \sqrt{2} \sin(2\pi f_c t + 135^\circ). \end{aligned}$$

$$\begin{aligned} \Rightarrow X_3 \rightarrow 10 &\rightarrow \sin 2\pi f_c t - \cos 2\pi f_c t \\ &= \sqrt{2} \sin(2\pi f_c t - 45^\circ) \end{aligned}$$

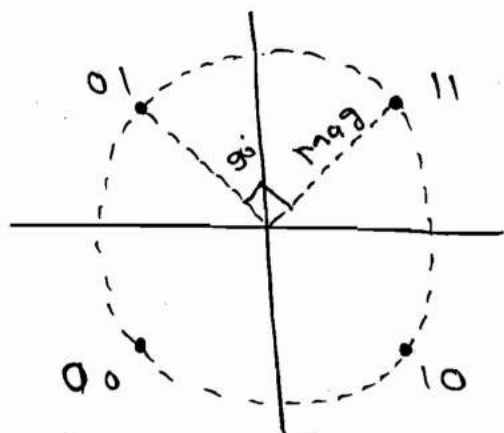
$$\begin{aligned} \Rightarrow X_4 \rightarrow 11 &\rightarrow \sin 2\pi f_c t + \cos 2\pi f_c t \\ &= \sqrt{2} \sin(2\pi f_c t + 45^\circ). \end{aligned}$$

* Phasor Diagram:

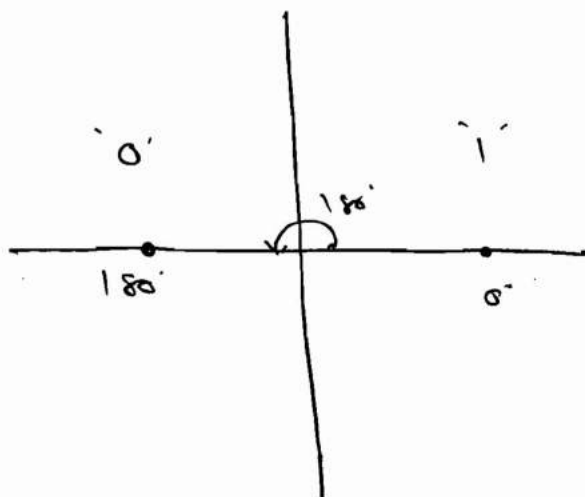


* Constellation Diagram:

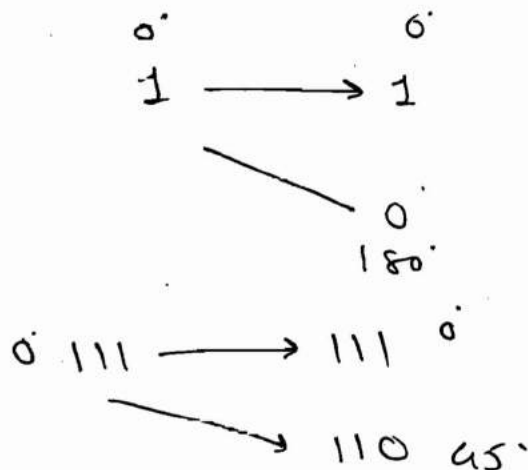
① QPSK:



② BPSK:

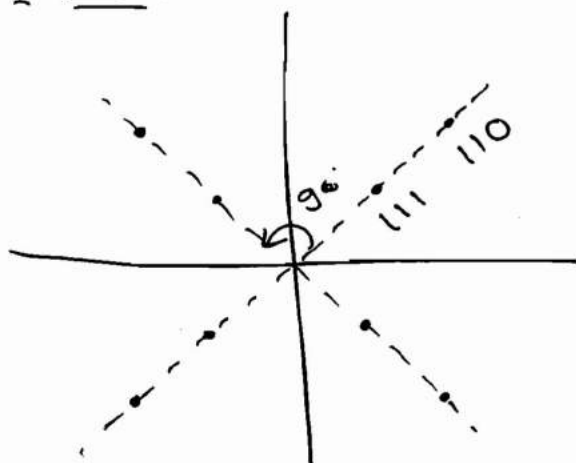


③ 8PSK:



⇒ As M increases angular suppression decreases
(or) probability of error increases.

④ 8QAM

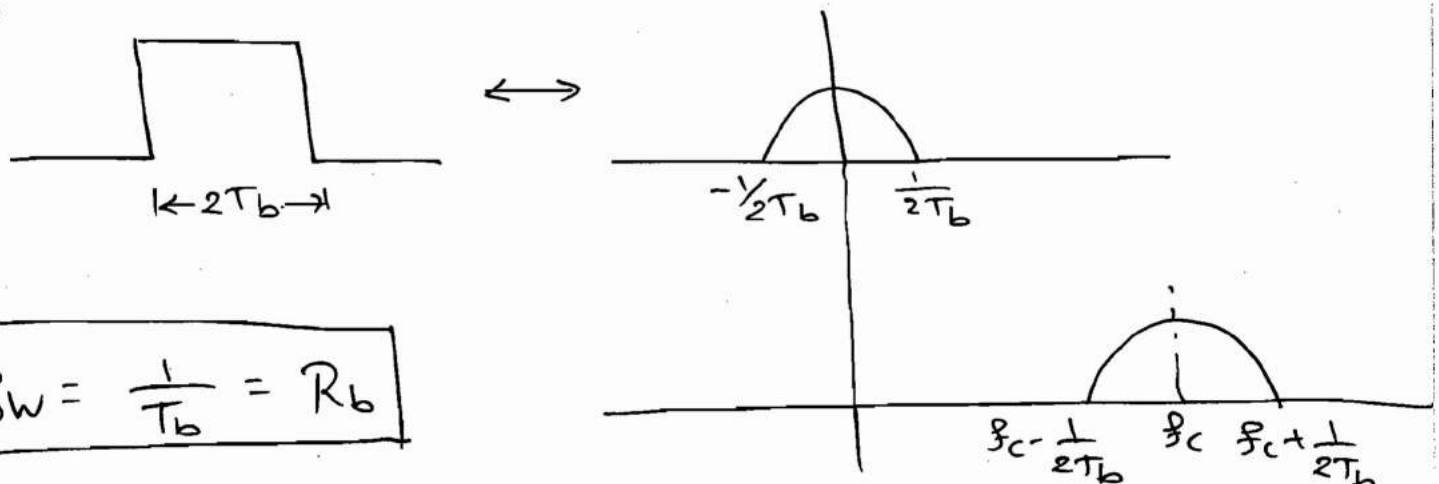


$\Rightarrow$ The input to the QPSK modulator is 2 bits at a time. The duration of this two bits is $2T_b$ sec. one bit is applied to the inphase channel and the other bit is applied to the quadrature channel. at the same time.

$\Rightarrow$ The input to the Balanced Modulator is a rectangular pulse but the pulse width is $2T_b$ sec.

$\Rightarrow$ one pulse is multiplied with the $\cos 2\pi f_c t$ and one pulse is multiplied by $\sin 2\pi f_c t$. So, both occupy the same freq. range but interference not occurs as the carriers are orthogonal to each other.

$\Rightarrow$



$$B_w = \frac{1}{T_b} = R_b$$

* Generalized formula to determine BW
of M-ary signaling:

⇒

$$BW = \frac{2R_b}{\log_2 M} \quad \underline{\text{IMP}}$$

① BPSK → $M=2$

$$BW = \frac{2R_b}{\log_2 2} = \underline{\underline{2R_b}}$$

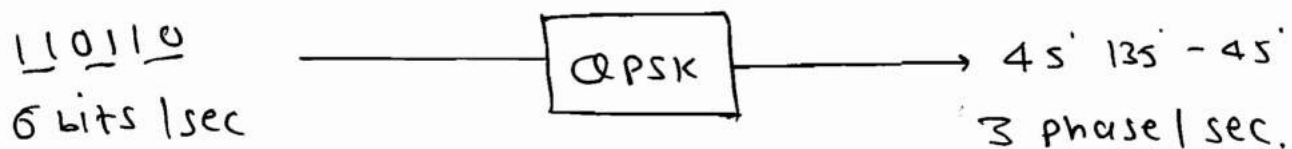
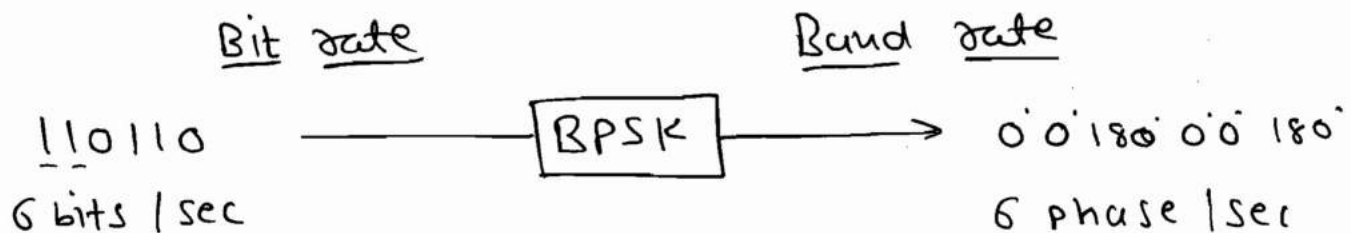
② QPSK → $M=4$
(QAM)

$$BW = \frac{2R_b}{\log_2 2^2} = \frac{2R_b}{2 \log_2 2} = \underline{\underline{R_b}}$$

③ 8-PSK → $M=8$

$$BW = \frac{2R_b}{\log_2 2^3} = \underline{\underline{\frac{2}{3} R_b}}$$

* Band rate:



$$\Rightarrow \therefore \text{Band rate} = \frac{\text{Bit rate}}{\log_2 M} = \frac{R_b}{\log_2 M} = \frac{BW}{2}$$

$\Rightarrow$ Band width efficiency :-

$$\eta = \frac{R_b}{BW}$$

$$BPSK \rightarrow \eta = \frac{R_b}{2R_b} = 50\%$$

$$QPSK \rightarrow \eta = \frac{R_b}{2R_b} = 100\%$$